



Chettinad

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Engineering

TOPIC: SATELLITE LINK DESIGN

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INTERFERENCE ANALYSIS

Key terms:

- ❖ Earth station → Associated with satellite circuits
- ❖ Terrestrial station → Associated with ground based microwave LoS circuits

Possible modes of Interference

- **A1:** Terrestrial station txion ($I \rightarrow ES$)
- **A2:** ES txion ($I \rightarrow TS$)
- **B1:** Space station txion of 1 space system \rightarrow ($I \rightarrow Rx$ by ES of another space system)
- **B2:** ES txion of 1 space system \rightarrow ($I \rightarrow Rx$ by space station (SS) of another space system)

- **C1:** SS txion(I → RX by a terrestrial station)
- **C2:** TS txion(I → Rx by a SS)
- **E:** SS txion of 1 space system(I → Rx SS by an another space system)

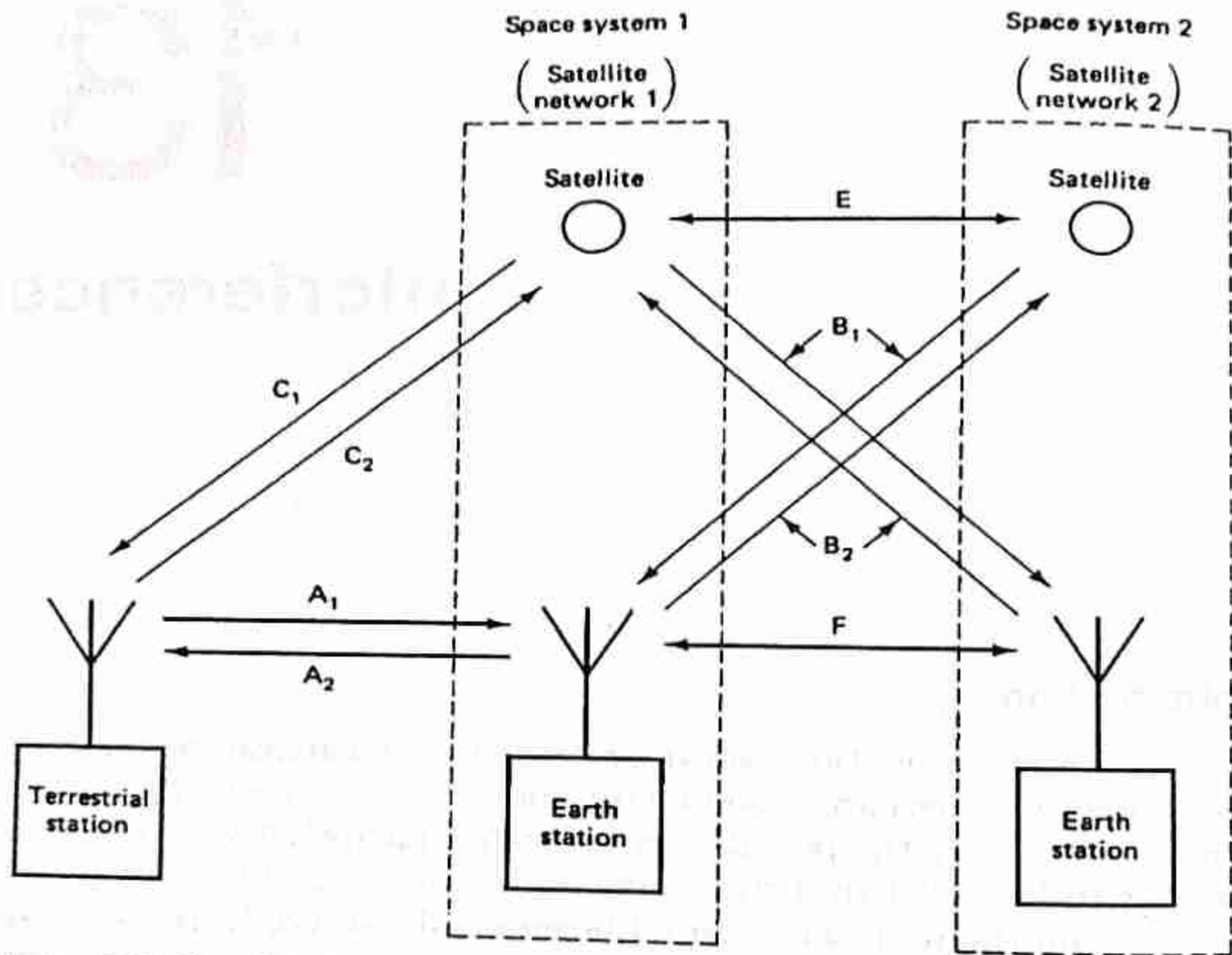


Figure 13.1 Possible interference modes between satellite circuits and a terrestrial station. (Courtesy of CCIR Radio Regulations.)

A_1 : terrestrial station transmissions, possibly causing interference to reception by an earth station

A_2 : earth station transmissions, possibly causing interference to reception by a terrestrial station

B_1 : space station transmission of one space system, possibly causing interference to reception by an earth station of another space system

B_2 : earth station transmissions of one space system, possibly causing interference to reception by a space station of another space system

C_1 : space station transmission, possibly causing interference to reception by a terrestrial station

C_2 : terrestrial station transmission, possibly causing interference to reception by a space station

E : space station transmission of one space system, possibly causing interference to reception by a space station of another space system

F : earth station transmission of one space system, possibly causing interference to reception by an earth station of another space system

- ❏ A1,A2,C1 and C2 → Possible modes of interference between space and terrestrial services
- ❏ B1 and B2 → Possible modes of interference between diff. Space system station
- ❏ E and F → extensions to B1 and B2

Interference between Satellite Circuits (B1 and B2 Modes)

- ◆ No. of Neighboring satellite circuits → aggregate interference
- ◆ 'T' considered as a form of noise → s/m performance is estimated by Interfering powers C/I ratio
- ◆ Imp. Parameter for controlling 'T' → RP of ES antenna

C/I must be related with ant. RP

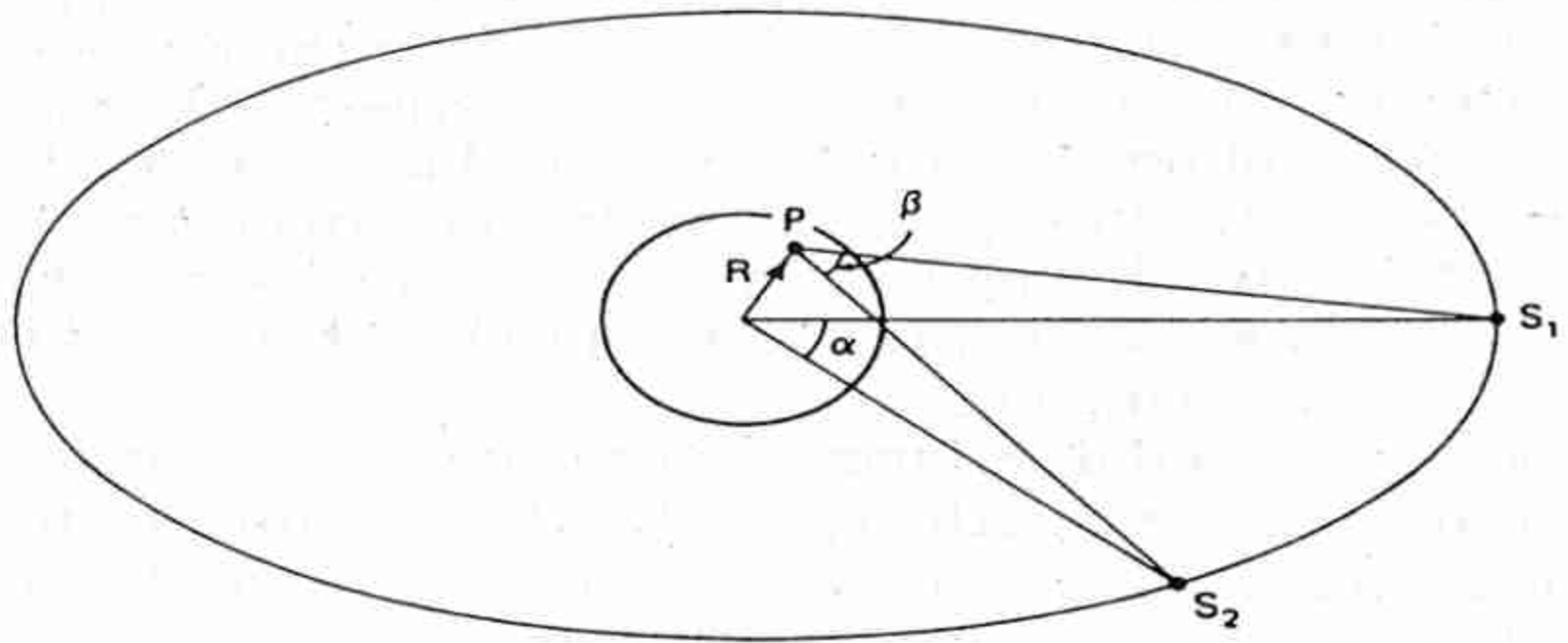


Figure 13.2 Geocentric angle α and the topocentric angle β .

- $\alpha \rightarrow$ Geocentric angle (subtended at the centre of the earth)
- $\beta \rightarrow$ Topocentric angle (from P \rightarrow the satellite would appear to subtend)
- Practically, $\alpha = \beta$
- S1 \rightarrow wanted sat
- S2 \rightarrow interfering sat

Orbital spacing angle

2° to 4° in 0.5° intervals in C band

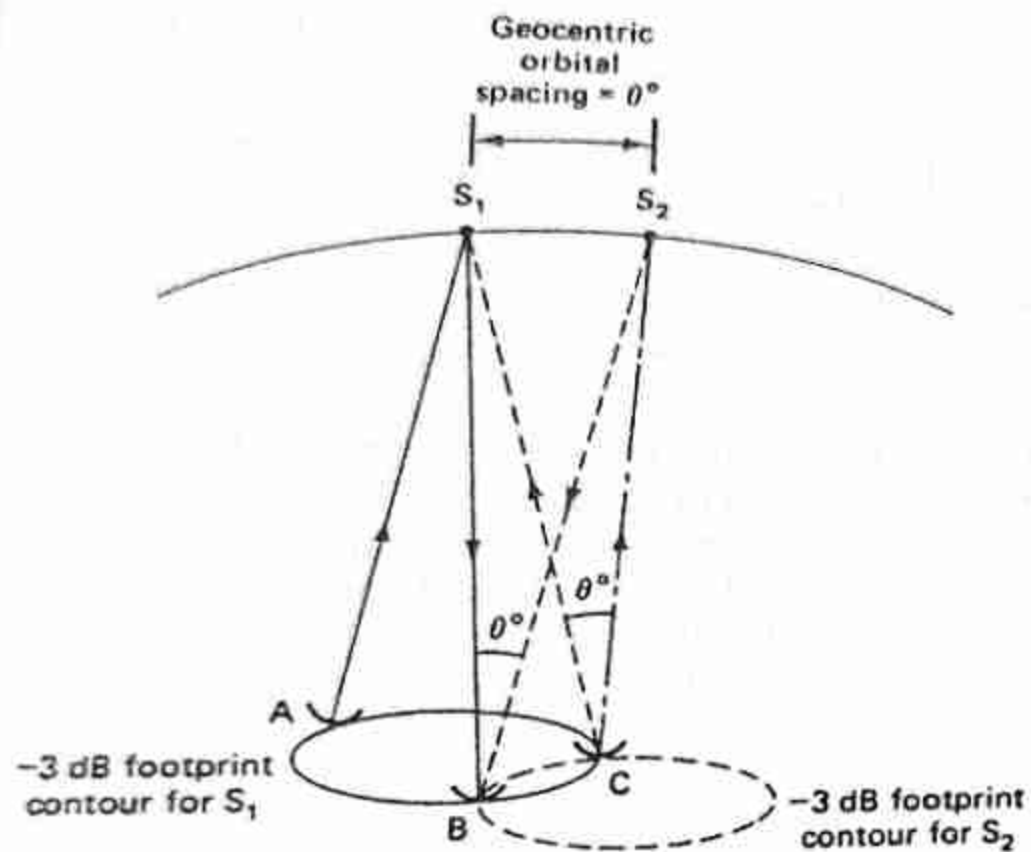


Figure 13.3 Orbital spacing angle.

Interference at sub-circ:

⇒ Neighbor sub-circ

⇒ Aggregate interference

⇒ c/I ratio

⇒ Important factor for controlling Interf: ⇒ "R.F" of ISM/1

⇒ large dia - Reflector in narrow BW

⇒ S_{ij} in-m ant: at 14.4MHz has a -3dB BW ant

0.15°

⇒ 'd' as substituted a size

⇓
orbital separation

(Geometric)

D/x:

$$[C] = [EIRP]_1 - 3 + [G_A] - [FSL]$$

$$[I] = [EIRP]_2 - 3 + [G_A(B)] - [FSL] - [Y]_D$$

↓
power discrimination

⇔

$$[C] - [I] = [EIRP]_1 - [EIRP]_2 + [G_A] - [G_A(B)] + [Y]_D$$

$$\left[\frac{C}{I} \right]_D = D [E] + [G_A] - [G_A(B)] + [Y]_D$$

U/x:

$$\left[\frac{C}{I} \right] = D [P] + [G_A] - [G_A(B)] + [Y]_D$$

↓
wanted & interfering
powers

↗ Beam sight
↘ off-axis
(focused main & collimated beam)

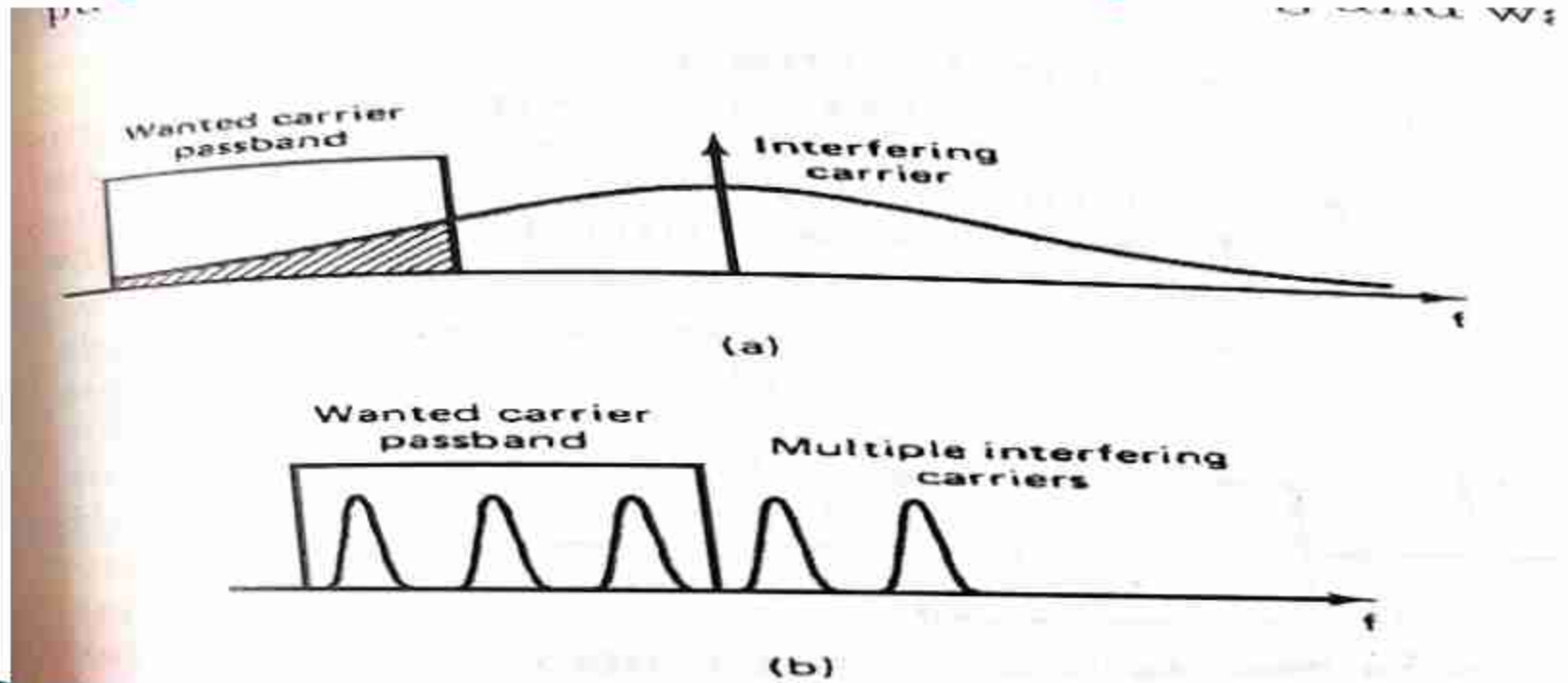
ANTENNA GAIN FUNCTION

- ✦ R.P → ÷ed into 3 regions
- ✦ 1. Mainlobe region
- ✦ 2. Sidelobe region
- ✦ 3. Transition region
- ✦ AGF defined the interference levels
- ✦ Off-axis angle = θ

$$[G(\theta)] = \begin{cases} 29 - 25 \log \theta & 1 \leq \theta \leq 7 \\ +8 & 7 < \theta \leq 9.2 \\ 32 - 25 \log \theta & 9.2 < \theta \leq 48 \\ -10 & 48 < \theta \leq 180 \end{cases}$$

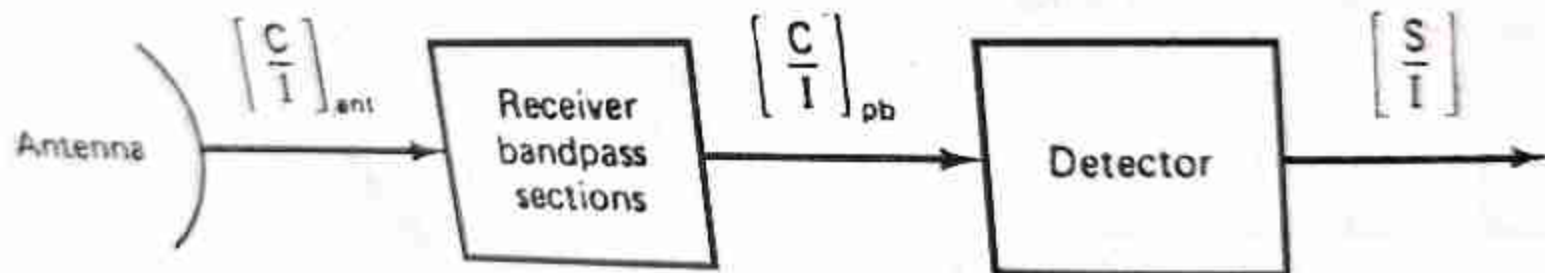
PASSBAND INTERFERENCE

“The amount of interference reaching the detector will depend on the amount of frequency overlap between the interfering spectrum and the wanted channel passband”



RECEIVER TRANSFER CHARACTERISTICS

- ❑ Baseband Interference
- ❑ $[S/I] = [C/I] + [RTC]$
- ❑ $RTC \rightarrow$ analogous



SPECIFIED INTERFERENCE OBJECTIVES

- ▶ For TV, viewing tests are conducted
- ▶ By gradually increasing the interference level, quality impairment factor can be established (1 to 5)

- ▶ **Protection Ratio** → “Minimum [C/I]
- ▶ **Energy dispersal** → Effective in reducing all modes of interference but particularly that occurring between ES and TS and also effective in reducing intermodulation noise
- ▶ **Interference levels**

$$[I_1] = [U_s] + [G'_s] + [G_E] - [L_D]$$

Rain induced attenuation and interference

- Rain attenuation is a major challenge to microwave satellite communication especially at frequencies above 10 GHz, causing unavailability of signals most of the time.
- Rain attenuation predictions have become one of the vital considerations while setting up a satellite communication link.

Rain induced attenuation and interference

“The attenuation of a signal due to rain is the extent to which the strength of a signal is reduced when passing through rain from a transmitter to a receiver, Also called rain fade.”

EFFECTS OF RAIN


- Rainfall causes attenuation of radio waves by scattering and by absorption of energy from the wave.
- Signal fading which affects the signal strength is caused mainly by rainfall

EFFECTS OF RAIN

Rain Attenuation:

- Rain attenuation increases with increasing frequency.
- It is worse in the Ku band than C band.
- Also, rain attenuation is accompanied by noise generation and both the attenuation and the noise affect the satellite circuit performance badly.

Rain Depolarization:

- When raindrops fall through the atmosphere, they get flattened in shape and become elliptical.
 - If the wave has some arbitrary polarization, it becomes elliptically polarized due to depolarization.
 - Depolarizing devices can be installed to compensate for rain depolarization in places where frequency reuse is achieved
- 

Effects on Radome

- Some earth station antennas are operated under cover of a radome.
- Rain falling on a hemispherical radome forms a water layer of constant thickness.
- This layer introduces losses by absorption and reflection.
- Results show that a 1-mm thick layer introduces an attenuation of 14 dB.
- Therefore, earth station antennas should be operated without radomes wherever possible.

Signal fading caused by rainfall places two limits which are:

- ▶ Uplink rain fade margin
- ▶ Downlink rain fade margin

UPLINK RAIN-FADE MARGIN

- Rainfall attenuates the signal and increases the noise temperature.
- This degrades the carrier to noise density ratio $[C/N_0]$ at the satellite in two ways:
 - 1. By increasing the noise
 - 2. By affecting power output

DOWNLINK RAIN FADE MARGIN

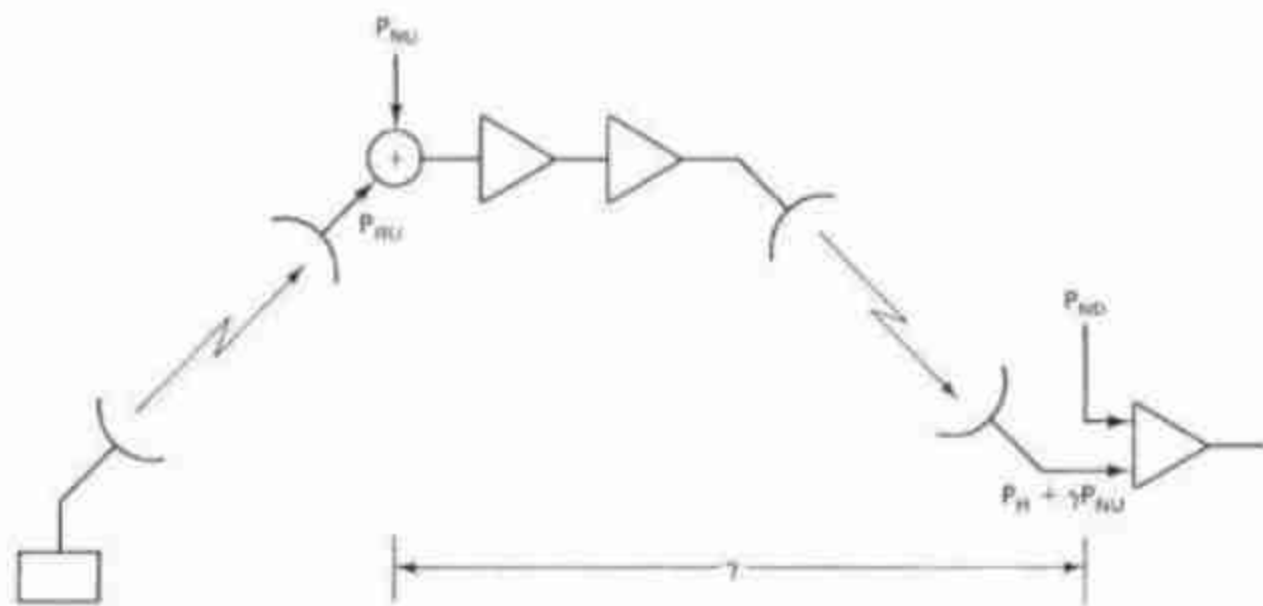
The Rainfall introduces attenuation by absorption and scattering of signal energy.

Therefore, the received carrier - to - noise density ratio $[C/N_0]$ is degraded by the rainfall in 2 ways.

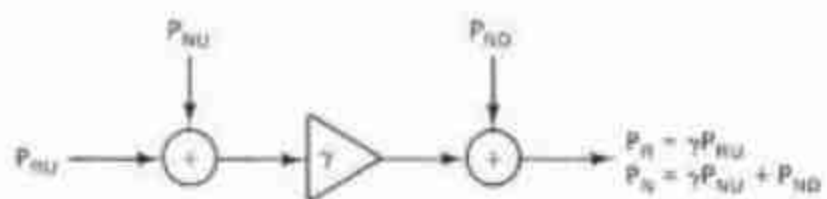
1. By attenuating the carrier wave
2. By increasing the sky noise temperature

COMBINED UPLINK AND DOWNLINK C/N RATIO

- Uplink rain-fade margin → Uplink power control for compensating rain fades
- Downlink rain fade margin → The C/No ratio for the downlink alone, not counting the P_{NU} contribution, is P_R/P_{ND} , and the combined C/No ratio at the ground receiver is



(a)



(b)

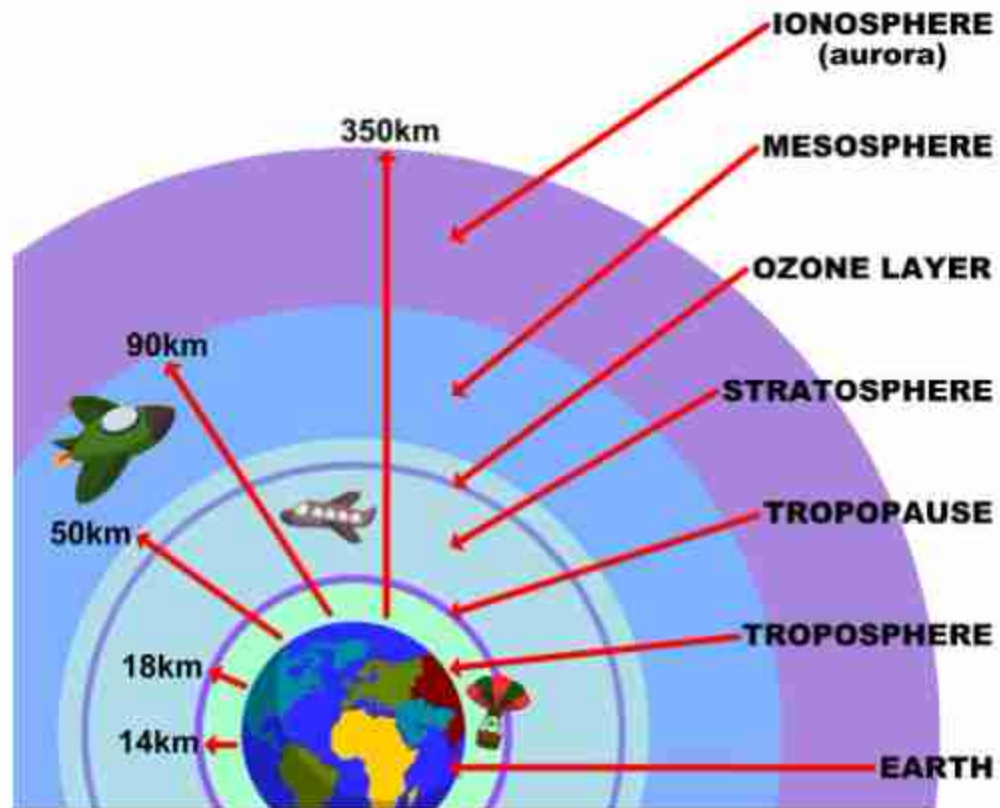
(a) Combined uplink and downlink; (b) power flow diagram

$$\begin{aligned}\frac{N_0}{C} &= \frac{P_N}{P_R} \\ &= \frac{\gamma P_{NU} + P_{ND}}{P_R} \\ &= \frac{\gamma P_{NU}}{P_R} + \frac{P_{ND}}{P_R} \\ &= \frac{\gamma P_{NU}}{\gamma P_{RU}} + \frac{P_{NU}}{P_R} \\ &= \left(\frac{N_0}{C}\right)_U + \left(\frac{N_0}{C}\right)_D\end{aligned}$$

- ▶ <https://www.youtube.com/watch?v=TmKnR8F2V2k>
- ▶ <https://www.youtube.com/watch?v=rcOY13FCoqY>

IONOSPHERIC CHARACTERISTICS

Layers of the Atmosphere



“The ionosphere is **where Earth's atmosphere meets space**”

- ▶ **Upper region** of Earth's atmosphere which has been ionized mainly by solar radiation
- ▶ This area of the atmosphere consists of several **conductive layers that reflect** radio waves
- ▶ Radio wave travelling between satellites and ES must pass through the ionosphere

- ▶ The free e^- in Ionosphere are **not uniformly distributed** but form in layers.
- ▶ Clouds of e^- may travel thru' it \rightarrow give rise to fluctuations in s/l \rightarrow only be determined on statistical basis

The **effects** include,

1. **Scintillation**
2. **Absorption**
3. **Variation in the direction of arrival**
4. **Propagation Delay**
5. **Dispersion**
6. **Frequency change**
7. **Polarization rotation**

Scintillation:

- ✦ Variations in amplitude, phase, polarization or angle of arrival of radio waves
- ✦ Caused by irregularities in ionosphere → change with time
- ✦ Main effect → Fading of signal → severe → necessary to include a fade margin in link power budget calculations

Absorption:

- 🔥 EM waves are absorbed in atm. According to W/L
- 🔥 2 Compounds are responsible for the majority of signal adsorption → O₂(63GHz) and Water(22GHz)

Propagation Delay

- ✎ “Time required for a signal to travel from S to D”

Dispersion:

- ✎ Signals are distributed over a wide area

Polarization rotation:

- ✎ Phenomenon in which waves of light or other radiation are restricted in direction of vibration.

Carrier to Noise Ratio: $\left[\frac{C}{N}\right]$ (or) CNR

"A measure of the performance of a satellite link"

⇒ Denoted by C/N

$$\frac{C}{N} = \frac{P_R}{P_N}$$

$$\left[\frac{C}{N}\right] = [P_R] - [P_N]$$

W.K.T,

$$[P_R] = [EIRP] + [G_R] - [\text{Losses}]$$

$$P_N = k T_s B_N$$

$$[P_N] = [k] + [T_s] + [B_N]$$

$$\therefore \left[\frac{C}{N}\right] = [EIRP] + [G_R] - [\text{Losses}] - [k] - [T_s] - [B_N]$$

⇒ $\frac{G}{T}$ ratio is a key parameter in specifying the receiving system performance.

$$\therefore [G/T] = [G_R] - [T_s] \text{ dBK}^{-1}$$

$$\therefore \left[\frac{C}{N}\right] = [EIRP] + \left[\frac{G}{T}\right] - [\text{Losses}] - [k] - [B_N]$$

$$\Rightarrow P_R = C, P_N = N$$

$$\therefore P_N = k T_N B_N = N_0 B_N$$

or
NPSD

$$\therefore \left[\frac{C}{N}\right] = \left[\frac{C}{N_0 B_N}\right]$$

$$= \left[\frac{C}{N_0}\right] - [B_N]$$

$$\therefore \left[\frac{C}{N_0}\right] = \left[\frac{C}{N}\right] + [B_N]$$

$$\left[\frac{C}{N_0} \right] = [EIRP] + \left[\frac{G}{T} \right] - [\text{Losses}] - [K]$$

In terms of decibels →

- | | | |
|------------------------|---|-------------|
| 1. FSL | → | -206 |
| 2. BAL | → | -2 |
| 3. APZ | → | -1 |
| 4. PL | → | -1 |
| 5. pol. Mismatch & | → | 0 |
| 6. Rx G/T | → | 19.5 |
| 7. EIRP | → | 48 |
| 8. - [K] | → | 20.6 |
| 9. [C/N ₀] | → | 86.1 ⇒ dBHz |

The Uplink: ('U' ⇒ used as subscript)

$$\left[\frac{C}{N_0} \right]_U = [EIRP]_U + \left[\frac{G}{T} \right]_U - [\text{Losses}]_U - [K]$$

⇒ Values to be used are the ES-EIRP, Feeder Loss, Rx G/T.

⇒ Saturation flux density:

"A measure of how much power a signal must have to saturate a detector"

⇒ "The flux density required at the receiving antenna to produce saturation of TWTB"

$$\Psi_{sat} = \frac{EIRP}{4\pi R^2}$$

In dB,

$$[\Psi_{sat}] = [EIRP] + 10 \log \frac{1}{4\pi R^2} \rightarrow \textcircled{1}$$

W.K.T \rightarrow

$$- [FSL] = 10 \log \frac{A^2}{4\pi r^2} + 10 \log \frac{1}{4\pi r^2}$$

$$\Rightarrow [\psi_M] = [EIRP] - [FSL] - 10 \log \left(\frac{A^2}{4\pi r^2} \right) \rightarrow \textcircled{2}$$

$\frac{A^2}{4\pi r^2} \Rightarrow$ Effective Area of an isotropic antenna.

$$[D_o] = 10 \log \left[\frac{A^2}{4\pi r^2} \right]$$

$$= 10 \log \left(\frac{(2 \times 10^8)^2}{4\pi \times (10^4)^2} \cdot \frac{1}{\beta^2} \right)$$

ψ_{loss}

$$= 10 \log \left(\frac{9 \times 10^{16}}{4\pi \times 10^8} \right) = 20 \log f$$

$$\approx 10 \log \left(\frac{9}{4\pi} \times 10^{-2} \right) - 20 \log f$$

$$= -21.45 - 20 \log f$$

$$= - (21.45 + 20 \log f)$$

$f = 50 \text{ GHz}$



$$\therefore \text{Eqn } \textcircled{2} \Rightarrow [EIRP] = [\psi_M] + [FSL] + [D_o]$$

Other losses,

$$[EIRP] = [\psi_M] + [D_o] + [FSL] + [COT] + [PL] + [AM]$$

$$[EIRP]_U = [\psi_S] + [D_o] + [Losses]_U - [RFL]$$

Input Backoff:

If a no. of carriers present simultaneously in a TWT, the operating point must be backed off to a linear portion of the transfer characteristics \downarrow the effects of intermodulation distortion. \Rightarrow occurs with FDMA. \Rightarrow Backoff must be considered.

$$[EIRP]_U = [EIRP_s]_U - [BO]_i$$

∴ W.K.T,

$$[EIRP_s]_U = [\psi_s] + [A_0] + [LOSSES]_U - [RFL]$$

$$\therefore \left[\frac{C}{N_0} \right]_U = [\psi_s] + [A_0] - [BO]_i + \left[\frac{G}{T} \right]_U - [K] - [RFL]$$

Downlink:

⇒ Sat. is transmitting the signal & the earth station is receiving it.

$$\therefore \left[\frac{C}{N_0} \right]_D = [EIRP]_D + \left[\frac{G}{T} \right]_D - [LOSSES]_D - [K]$$

W.K.T

$$P_N = kT_N B_N$$

$$kT_N = N_0$$

$$\therefore \left[\frac{C}{N} \right]_D = [EIRP] + \left[\frac{G}{T} \right]_D - [LOSSES]_D - [K] - [B]$$

THANK YOU

